April 2, 2019

# Introduction of Surface Acoustic Wave (SAW) Devices

Part 6: 2D Propagation and Waveguide

Ken-ya Hashimoto

Chiba University

k.hashimoto@ieee.org

http://www.te.chiba-u.jp/~ken

# Contents

- Wavevector and Diffraction
- Waveguide
- Scalar Potential Theory

# Contents

• Wavevector and Diffraction



Critical Length :  $x_c = (1+\gamma)W^2/\lambda$ 

γ: Parameter Determined by Anisotropy(=0 for Isotropic)



Variation with Aperture Size



### For Weighted IDT

# Wave Vector **B**

 $|\beta|=2\pi/\lambda$ :Phase Delay per Unit Length



 $V_{\rm p}$  (= $f\lambda$ ) does not Follow Vector Decomposition Rule!

 $\exp(-j\boldsymbol{\beta} \bullet \mathbf{X}) \implies \exp[-j(\boldsymbol{\beta}_x x + \boldsymbol{\beta}_y y + \boldsymbol{\beta}_z z)]$ 



# Snell's Law

### **Continuity of Wave Front at Boundary**



### Continuity of Lateral Wavelength ⇒ Continuity of Lateral Wavevector Component

### At Boundary Between Two media,



## **Evanescent Field (at Total Reflection)**

$$\beta_{y}^{(2)} = -j\sqrt{\beta_{x}^{2} - \beta_{0}^{(2)2}}$$



**Field Penetration** 

Exponential Decay (Energy Storage)





- Even for Total Reflection State, Wave Transmission Occurs when Medium is Thin
- No Phase Delay Through Transmission

**Anisotropy Case**  $u \propto \exp[j(\omega t - \beta_x x - \beta_y y)]$ 

$$\beta_x^2 + (V_{y0} / V_{x0})^2 \beta_y^2 = (\omega / V_{x0})^2$$
$$\beta_x \cong (\omega / V_{x0}) - 2^{-1} (\omega / V_{x0})^{-1} (V_{y0} / V_{x0})^2 \beta_y^2$$



Parabolic Approximation





### Para-Axial Approximation for X»|Y|

Approximating  $\beta_x \cong \beta_{x0} - \zeta \beta_y^2$ , Then  $G(X,Y) \cong \frac{F}{\sqrt{2\pi |\zeta X|}} \exp(-j\beta |X| - jY^2 / 4 |\zeta X|)$  Contribution of *n*-th Electrode (Width  $w_n$ , Position  $(x_n, y_n)$ )







#### Significant at Higher Out-of-Band Rejection

# Contents

• Waveguide

## Influence of Diffraction in SAW Resonators





## **Inharmonic Resonances**



Design Challenge: Suppression of Inharmonic Resonances Without Badly Affecting Main Resonance



For Phase Matching Between Incident and 2-Bounced Waves

*Transverse Resonance Condition*  $-2\beta_v h_v + 2\angle \Gamma = 2n\pi$ 



 $\beta_v h_v = n_v \pi$ 

 $\beta_v h_v = n_v \pi$ 

 $\left(\frac{n_{y}\pi}{h_{v}}\right)^{2} = \beta_{0}^{2}$ 

 $\beta_x^2 + \left(\frac{n_y \pi}{h_v}\right)^2 = \beta_0^2$ 

$$\beta_x = \sqrt{\beta_0^2 - \beta_y^2}$$
 and  $\beta_y h_y = n\pi$   
  
Wavenumber of Guided Mode  $\beta_x = \sqrt{(\omega/V)^2 - (n\pi/h_y)^2}$ 



Relation Between  $\beta_x$  and  $\beta_0$  When  $\angle \Gamma = 0$  or  $\pm \pi$ 



## **Influence of Group and Phase Velocities on Signal Transfer**



### **Under Cutoff Frequency**





**Behavior as Evanescent Field** 

Even if not Cutoff



Influence of Higher-Order Cutoff Modes

# **Open Waveguide**



Use of Total Reflection at Surfaces  $\Rightarrow$ Energy Penetration to Outsides

*Transverse Resonance Condition*  $-2\beta_y h_y + 2\angle \Gamma = 2n\pi$  $\angle \Gamma$  is Frequency (or  $\theta$ ) dependent

### Similarity with Closed Waveguide at Total Reflection



Relation between  $\beta_x$  and  $\beta_0$ 

If Total Reflection Condition is Not Satisfied?

## Leaky Waveguide



When Reflection Coefficient at Surfaces is Large, Pseudo Mode Propagates with Energy Leakage to Outside

If Reflection Coefficient at Surfaces is Small?

### **Propagation as Free Wave(Not Guided)**



Appearing When Velocities of Waveguide Mode and Free Wave are Close (Near Cutoff)



 $\phi_{\rm c} = \cos^{-1}(V_{\rm S}/V_{\rm B})$ : critical angle

 $V_{\rm S}$ : SAW velocity,  $V_{\rm B}$ : BAW velocity





 $\phi_{\rm c} = \cos^{-1}(V_{\rm B}/V_{\rm S})$ : critical angle

*V*<sub>S</sub>: SAW velocity, *V*<sub>B</sub>: BAW velocity *Field Amplitude Grows Toward the Depth!*  **Resonance Frequency of Cuboid Cavity** 



Wavevector of Propagation Mode in Rectangular Waveguide





• Scalar Potential Theory

## **Scalar Potential Analysis**



Due to Continuity of  $\phi$  and  $\partial \phi / \partial y$  at  $y = \pm w_G / 2$ 

Symmetric Mode  $(\phi_B^+ = \phi_B^-, \phi_G^+ = \phi_G^-)$   $\phi_B^+ = 2\phi_G^+ \cos(\beta_{Gy} w_G / 2) \exp(\alpha_{By} w_G / 2)$  $\alpha_{By} = \beta_{Gy} \tan(\beta_{Gy} w_G / 2)$ 

Anti-Symmetric Mode  $(\phi_B^+ = -\phi_B^-, \phi_G^+ = -\phi_G^-)$   $\phi_B^+ = -2j\phi_G^+ \sin(\beta_{Gy}w_G/2)\exp(\alpha_{By}w_G/2)$  $\alpha_{By} = -\beta_{Gy}\cot(\beta_{Gy}w_G/2)$ 

### **Parabolic Approximation for Slowness Surface**



For Region G  $\beta_x \cong \beta_{G0} - \xi_G \beta_{Gy}^2 / \beta_{G0}$ For Region B  $\beta_x \cong \beta_{B0} + \xi_B \alpha_{By}^2 / \beta_{B0}$ For Isotropic Case,  $\xi=0.5$ 

## Slowness Surface of SH-type SAW on 36-LT



### **Wavenumber of Grating Mode and Slowness Surface**

For Energy Trapping in Waveguide  $\Rightarrow$  Real  $\alpha_{Bv}$ 



When  $|V_{B0}/V_{G0}-1| \ll 1$ , Symmetric Mode  $\sqrt{\frac{1-\hat{V}}{1+\hat{V}}} = \tan\left[\pi\hat{w}_{G}\sqrt{\frac{1+\hat{V}}{2}}\right]$ Anti-Symmetric Mode  $\sqrt{\frac{1-\hat{V}}{1+\hat{V}}} = -\cot\left[\pi\hat{w}_{G}\sqrt{\frac{1+\hat{V}}{2}}\right]$ 

Where 
$$\hat{V} = \frac{2V_p - V_{B0} - V_{G0}}{V_{B0} - V_{G0}}$$
 :Relative Phase Velocity  
 $\hat{W}_{G} = \frac{W_G}{\lambda_p} \sqrt{\frac{V_{B0} - V_{G0}}{\xi V_{G0}}}$  :Relative Waveguide Width

**Relative SAW Velocity vs. Relative Aperture** 



Velocity in Region B ⇒Velocity in Region G

### **Equivalent Circuit for Multi-Mode Resonators**



$$\omega_{\rm r}^{(n)} = \frac{1}{\sqrt{C_{\rm m}^{(n)}L_{\rm m}^{(n)}}} = \frac{2\pi V_{\rm p}^{(n)}}{p_{\rm I}}$$

### Modes Propagate without Mutual Power Interaction

$$\int_{-\infty}^{+\infty} |\phi_k(y) + \phi_n(y)|^2 dy = \int_{-\infty}^{+\infty} |\phi_k(y)|^2 dy + \int_{-\infty}^{+\infty} |\phi_n(y)|^2 dy$$

Mode Orthogonality

$$\int_{-\infty}^{+\infty} \phi_k(y) \phi_n^*(y) dy = \delta_{nk} P_k$$
  
where  $P_k = \int_{-\infty}^{+\infty} |\phi_k(y)|^2 dy$ 

**Field can be Expressed as Sum of Mode Fields** Mode Completeness

$$\phi(y) = \sum_{k=1}^{\infty} A_k \phi_k(y) / \sqrt{P_k}$$

## Fourier Transform $\varphi_n(x) = p^{-0.5} exp(2n\pi j x/p)$

Orthogonality

$$\int_{0}^{p} \phi_{k}(x)\phi_{n}^{*}(x)dx = p^{-0.5}\int_{0}^{p} \exp[2\pi j x(n-m)/p]dx = \delta_{nk}$$

$$\phi(x) = \sum_{k=1}^{\infty} A_k \phi_k(x) = p^{-0.5} \sum_{k=1}^{\infty} A_k \exp(2k\pi j x / p)$$

 $\begin{aligned} & \underset{p}{\overset{p}{\int}} \phi(x)\phi_{n}^{*}(x)dx = \int_{0}^{p} \sum_{k=1}^{\infty} A_{k}\phi_{k}(x)\phi_{n}^{*}(x)dx = A_{n} \\ & \underset{p}{\overset{p}{\int}} A_{n} = p^{-0.5} \int_{0}^{p} \phi(x) \exp(-2n\pi j x / p) dx \end{aligned}$ 

## **Difference of Waveguide Width** $w_g$ with Finger Overlap Width $w_e$



Amplitude at Excitation Source

$$\phi(y) = \begin{cases} \phi_0 & (|y| \le w_e/2) \\ 0 & (|y| > w_e/2) \end{cases}$$

### Multiplying $\phi_m^*(y)$ and Integrating

$$\int_{-\infty}^{+\infty} \phi(y) \phi_m^*(y) dy = \sum_{k=1}^{\infty} A_k / \sqrt{P_k} \int_{-\infty}^{+\infty} \phi_k(y) \phi_m^*(y) dy$$

Then

$$A_{m} = \frac{\phi_{0}}{\sqrt{P_{m}}} \int_{-w_{e}/2}^{+w_{e}/2} \phi_{m}^{*}(y) dy$$

1D Analysis Gives  $A_0 = \phi_0 \sqrt{w_e}$ 

Since Motional Capacitance  $\infty$  Power Excitation Efficiency,

$$\frac{C_m^{(n)}}{C_m^{(0)}} = \left|\frac{A_n}{A_0}\right|^2 = \left|\int_{-w_e/2}^{+w_e/2} \phi_n(y) dy\right|^2 \left[w_e \int_{-\infty}^{+\infty} |\phi_n(y)|^2 dy\right]^{-1}$$

### Effective Electromechanical Coupling Factor vs. Relative Aperture Width (When $w_e = w_g$ )



**Zero Excitation Efficiency for Anti-Symmetric Modes** 

## **Why Effective Coupling Factor Changes?**

